## Introduction

The FLEX ${ }^{\circledR}$ 10K and FLEX 8000 device families can efficiently implement two-dimensional convolvers, which are useful for many image processing applications such as filtering, sharpening, edge detection, and interpolation. A convolver processes information in small pieces, such as a $3 \times 3$ convolution window. Each element in the convolution window is multiplied by a constant coefficient, and then the products are summed to obtain the desired result.

This application note describes how to create more efficient convolvers in FLEX 10K and FLEX 8000 designs.

For more information on two-dimensional convolvers, see the following documents:

## - Application Note 73 (Implementing FIR Filters in FLEX Devices) <br> - Product Information Bulletin 21 (Implementing Logic with the Embedded Array in FLEX 10K Devices)

For this application note, the 9 locations in a $3 \times 3$ convolution window are numbered for reference, as shown in Figure 1. The subscript indicates which tap or coefficient is being referenced. The variable $T_{n}$ refers to the tap data value at location $n$ (e.g., $T_{2}$ refers to the tap value at location 2 ).

Figure 1. Tap \& Coefficient Reference Locations

| 1 | 2 | 3 |
| :--- | :--- | :--- |
| 4 | 5 | 6 |
| 7 | 8 | 9 |

Conventional Convolver

Figure 2 shows the coefficient window for a sample convolver design.

Figure 2. Coefficient Window

| Row 1 | 1.5 | 0.375 | -0.625 |
| :--- | :---: | :---: | :---: |
| Row 2 | 0.375 | 0.125 | -0.125 |
| Row 3 | 1.25 | -0.125 | 0.250 |
|  |  |  |  |

Figure 3 shows a block diagram of a conventional convolver using the coefficients shown in Figure 2. In this convolver, the coefficients remain constant and the tap values change every clock cycle as new data is shifted into the convolver.

Figure 3. Conventional Convolver Block Diagram


The output of each register is called a tap $\left(T_{n}\right)$. Each tap is multiplied by the corresponding coefficient $\left(C_{n}\right)$ from Figure 2, and then the products are summed to produce the final result $(Y)$. The equation for the convolver in Figure 3 is shown below:
$Y=\sum_{1}^{9} T_{n} C_{n}$

Building the conventional two-dimensional convolver for maximum speed requires 9 multipliers and an adder tree consisting of 8 separate adders. Assuming the data and coefficients are 8 bits wide and the coefficients are constant, each multiplier would require 36 logic elements (LEs) and each adder would require one LE per bit, for a total of 539 LEs for the entire design.

Highly Optimized Convolver

You can improve the speed and logic efficiency of the conventional convolver by taking advantage of the numeric relationships between the coefficients. Although the convolution window in Figure 2 is asymmetric, all of the coefficients are integer multiples of 0.125 . By rewriting the coefficient window and substituting $K=0.125$, you obtain the window shown in Figure 4.

Figure 4. Modified Coefficient Window

| 12 | 3 | -5 |
| :---: | :---: | :---: |
| 3 | 1 | -1 |
| 10 | -1 | 2 |

In this example, you can take advantage of the fact that all the factored coefficients are integers. Multiplying the taps by the coefficients $12,3,5,1$, and 10 is simple because all the integers are the sum of powers of 2 . For example, multiplying tap $T_{1}$ by 12 is the same as adding $8 T_{1}$ and $4 T_{1}$ (i.e., $12 T_{1}=8 T_{1}+4 T_{1}$ ). Because 8 and 4 are both powers of 2 , multiplication can be avoided. Instead, $T_{1}$ can be shifted left by 3 bits and 2 bits, respectively. Therefore, $12 T_{1}$ can be computed with a single adder that is only 10 bits wide, assuming $T_{1}$ is an 8-bit number. For example, if $T_{1}=11111111_{\mathrm{B}}$, you can calculate $12 T_{1}$ as shown in Figure 5.

Figure 5. Multiplying $12 \times T_{1}$

$$
\begin{array}{r}
111111111000 \\
+1111111100 \\
\hline 101111110100
\end{array}=\quad \begin{array}{r}
8 \times T_{1} \\
+4 \times T_{1} \\
\hline 12 \times T_{1}
\end{array}
$$

Because the 2 least significant bits (LSBs) of $8 T_{1}$ and $4 T_{1}$ are always 0 , the 2 extra bits in the adder are not required. Therefore, this calculation requires only 10 LEs.

Similarly, 3, 5, and 10 can each be computed with a single adder. Multiplying or dividing by any power of 2 is a shift left or right and does not require any LEs. However, LEs are required to add the product into the final result.

To further improve the logic efficiency of this design, you can take advantage of the "sparseness" of the coeffiecients (i.e., the relatively few 1 s in the coefficients). The basic method is to add or subtract various multiples of 2 of the taps to obtain the desired multiplication factor-e.g., add together all taps that have a 1 in the most significant bit (MSB). Table 1 lists each tap's coefficient and its binary representation. Numbers in parentheses are negative.

| Table 1. Binary Representation of the 9 Coefficients |  |  |
| :---: | :---: | :---: |
| Coefficient | Weight |  |
|  | Decimal | Binary |
| $C_{1}$ | 12 | 1100 |
| $C_{2}$ | 3 | 0011 |
| $C_{3}$ | -5 | $(0101)$ |
| $C_{4}$ | 3 | 0011 |
| $C_{5}$ | 1 | 0001 |
| $C_{6}$ | -1 | $(0001)$ |
| $C_{7}$ | 10 | 1010 |
| $C_{8}$ | -1 | $(0001)$ |
| $C_{9}$ | 2 | 0010 |

To perform these calculations efficiently in parallel, begin by adding taps with a 1 in the MSB and work your way down to the LSB. In this example, the MSB of $C_{1}$ and $C_{7}$ is 1 . Therefore, you should calculate an intermediate result by adding $C_{1}$ and $C_{7}$, remembering to shift the product left by 3 bits before adding it to the final result.

Similarly, the (MSB - 1) bit of $C_{1}$ and $C_{3}$ is 1 . In this case, compute $\left(C_{1}-C_{3}\right)$, remembering to shift the product left by 2 bits before adding it to the final result.

Repeat this procedure for the next 2 bits of the weight, i.e., the (MSB - 2) bit and the LSB. Next, add all of the products to produce the final result. Last, multiply the final result by K. Because 0.125 is a power of 2 , no multiplication is necessary-the final result is simply shifted right by 3 bits.

The following steps show mathematically how to use this method to create a highly optimized convolver design, or any other multiply and accumulate (MAC) design. All numbers are in binary.

$$
Y=T_{1} C_{1}+T_{2} C_{2}+T_{3} C_{3}+T_{4} C_{4}+T_{5} C_{5}+T_{6} C_{6}+T_{7} C_{7}+T_{8} C_{8}+T_{9} C_{9}
$$

1. Substitute the binary values of the coefficients:

$$
\begin{aligned}
Y= & \left(T_{1} \times 1100\right)+\left(T_{2} \times 0011\right)-\left(T_{3} \times 0101\right)+\left(T_{4} \times 0011\right)+ \\
& \left(T_{5} \times 0001\right)-\left(T_{6} \times 0001\right)+\left(T_{7} \times 1010\right)-\left(T_{8} \times 0001\right)+\left(T_{9} \times 0010\right)
\end{aligned}
$$

2. Factor out $8\left(1000_{\mathrm{B}}\right)$ :

$$
\begin{aligned}
Y= & {\left[\left(T_{1}+T_{7}\right) \times 1000\right]+\left(T_{1} \times 100\right)+\left(T_{2} \times 011\right)-\left(T_{3} \times 101\right)+} \\
& \left(T_{4} \times 011\right)+\left(T_{5} \times 001\right)-\left(T_{6} \times 001\right)+\left(T_{7} \times 010\right)-\left(T_{8} \times 001\right)+\left(T_{9} \times 010\right)
\end{aligned}
$$

3. Factor out $4\left(100_{\mathrm{B}}\right)$ :

$$
\begin{aligned}
Y= & {\left[\left(T_{1}+T_{7}\right) \times 1000\right]+\left[\left(T_{1}-T_{3}\right) \times 100\right]+\left(T_{1} \times 00\right)+\left(T_{2} \times 11\right)-} \\
& \left(T_{3} \times 01\right)+\left(T_{4} \times 11\right)+\left(T_{5} \times 01\right)-\left(T_{6} \times 01\right)+\left(T_{7} \times 10\right)-\left(T_{8} \times 01\right)+ \\
& \left(T_{9} \times 10\right)
\end{aligned}
$$

4. Factor out $2\left(10_{\mathrm{B}}\right)$ :

$$
\begin{aligned}
Y= & {\left[\left(T_{1}+T_{7}\right) \times 1000\right]+\left[\left(T_{1}-T_{3}\right) \times 100\right]+\left[\left(T_{2}+T_{4}+T_{7}+T_{9}\right) \times 10\right]+} \\
& \left(T_{1} \times 0\right)+\left(T_{2} \times 1\right)-\left(T_{3} \times 1\right)+\left(T_{4} \times 1\right)+\left(T_{5} \times 1\right)-\left(T_{6} \times 1\right)+\left(T_{7} \times 0\right)- \\
& \left(T_{8} \times 1\right)+\left(T_{9} \times 0\right)
\end{aligned}
$$

5. Add the remaining non-zero terms:

$$
\begin{aligned}
Y= & {\left[\left(T_{1}+T_{7}\right) \times 1000\right]+\left[\left(T_{1}-T_{3}\right) \times 100\right]+\left[\left(T_{2}+T_{4}+T_{7}+T_{9}\right) \times 10\right]+} \\
& {\left[\left(T_{2}-T_{3}+T_{4}+T_{5}-T_{6}-T_{8}\right)\right] \times 1 }
\end{aligned}
$$

6. Group the terms by significance and perform more factoring:

$$
\begin{aligned}
Y= & {\left[\left(T_{1}+T_{7}\right) \times 10\right]+\left[\left(T_{1}-T_{3}\right) \times 1\right] \times 10+\left[\left(T_{2}+T_{4}+T_{7}+T_{9}\right) \times 1 \times 10\right]+} \\
& {\left[\left(T_{2}-T_{3}+T_{4}+T_{5}-T_{6}-T_{8}\right) \times 1\right] }
\end{aligned}
$$

Figure 6 shows a block diagram of the multiply and accumulate section of the optimized convolver.

Figure 6. Highly Optimized Convolver Block Diagram


The highly optimized fixed-coefficient convolver uses 215 LEs, as opposed to the 529 LEs required by the conventional convolver shown in Figure 3-a savings of $60 \%$.

The highly optimized design, written in the Altera Hardware Description Language ( $\mathrm{AHDL}^{\mathrm{TM}}$ ), is shown in Figure 7. This pipelined design will run at over 90 MHz in a FLEX 8000A device with a -2 speed grade, a speed that is the equivalent of 810 MIPS for only 214 LEs.
[1 This design-named fstconv-can be downloaded from the Altera world-wide web site at http://www.altera.com.

## Figure 7. fstconv. tdf File (Part 1 of 2)

SUBDESIGN fstconv
(
in[3..1][8..1] : INPUT;
out[15..1] : OUTPUT;
clk : INPUT;

## )

## VARIABLE

```
row1[3..1][8..1] : DFF; -- Row 1 of shift register
row2[3..1][8..1] : DFF; -- Row 2 of shift register
row3[3..1][8..1] : DFF; -- Row 3 of shift register
inter_sh3[9..1] : DFF; -- Intermediate shift by 3 node
inter_sh2[9..1] : DFF; -- Intermediate shift by 2 node
inter_sh1_t[2..1][9..1] : DFF; -- Shift by 1 node
inter_sh1[10..1] : DFF; -- More shift by 1 node
inter_sh0_t1[3..1][9..1]: DFF; -- Shift by 0
inter_sh0_t2a[10..1] : DFF;
inter_sh0_t2b[9..1] : DFF;
inter_sh0[11..1] : DFF;
inter_3_2[11..1] : DFF; -- Shift by 3 + shift by 2
inter_3_2_1[13..1] : DFF; -- inter_3_2 + shift by 1
inter_3_2_1_0[15..1] : DFF; -- inter_3_2_1 + shift by 0
```

BEGIN
-- Connect the clock to all the internal registers
(inter_sh3[].clk, inter_sh2[].clk, inter_sh1_t[][].clk, inter_sh1[].clk, inter_sh0_t1[][].clk, inter_sh0_t2a[].clk, inter_sh0_t2b[].clk, inter_sh0[11..1].clk, inter_3_2[].clk, inter_3_2_1[].clk inter_3_2_1_0[].clk) = clk;
-- Connect the shift register:
-- Data will shift from left to right, which will be 1 to 3
row1[1][] = in[1][];
row2[1][] = in[2][];
row3[1][] = in[3][];
row1[3..2][] = row1[2..1][];
row2[3..2][] = row2[2..1][];
row3[3..2][] = row3[2..1][];
row1[][].clk = clk;
row2[][].clk = clk;
row3[][].clk = clk;

Figure 7. fstconv. tdf (Part 2 of 2)
-- The shift by 3 data

```
inter_sh3[] = (row1[1][8], row1[1][]) + (row3[1][8], row3[1][]);
```

-- The shift by 2 data

```
inter_sh2[] = (row1[1][8], row1[1][]) - (row1[3][8], row1[3][]);
```

-- The shift by 1 data

```
inter_sh1_t[1][] = (row1[2][8], row1[2][]) +
    (row2[1][8], row2[1][]); -- 2 + 4
inter_sh1_t[2][] = (row3[1][8], row3[1][]) +
    (row3[3][8], row3[3][]); -- 7 + 9
inter_sh1[] = (inter_sh1_t[1][9], inter_sh1_t[1][]) +
    (inter_sh1_t[2][9], inter_sh1_t[2][]);
```

-- The shift by zero data

```
inter_sh0_t1[1][] = (row1[2][8], row1[2][]) +
                                    (row2[1][8], row2[1][]);
inter_sh0_t1[2][] = (row2[2][8], row2[2][]) -
                                    (row2[3][8], row2[3][]);
inter_sh0_t1[3][] = (row1[3][8], row1[3][]) +
                                    (row3[2][8], row3[2][]);
inter_sh0_t2a[] = (inter_sh0_t1[1][9], inter_sh0_t1[1][]) +
                                    (inter_sh0_t1[2][9], inter_sh0_t1[2][]);
inter_sh0_t2b[] = inter_sh0_t1[3][];
inter_sh0[] = (inter_sh0_t2a[10], inter_sh0_t2a[]) -
    (inter_sh0_t2b[9], inter_sh0_t2b[9],
    inter_sh0_t2b[]);
```

-- Shift by $3+$ shift by 2

```
inter_3_2[] = (inter_sh3[9], inter_sh3[], gnd) +
    (inter_sh2[9], inter_sh2[9], inter_sh2[]);
```

-- Shift by $3+$ shift by $2+$ shift by 1

```
    inter_3_2_1[] = (inter_3_2[11], inter_3_2[], gnd) +
    (inter_sh1[10], inter_sh1[10], inter_sh1[10],
    inter_sh1[]);
```

-- Shift by $3+2+1+0$
inter_3_2_1_0[] $=$ (inter_3_2_1[13], inter_3_2_1[], gnd) +
(inter_sh0[11], inter_sh0[11],
inter_sh0[11], inter_sh0[11], inter_sh0[]);
-- Output
out [] = inter_3_2_1_0[];
END;

## Conclusion

This application note shows how to improve the logic efficiency of a convolver design when the coefficients can be factored and are sparse. To reduce the amount of logic required for your convolver design, use the following tips:

- If possible, factor out a common coefficient, so that a single final multiplication can be performed at the end (if the factor is a multiple of 2 , multiplication is not required).
- If possible, set all the coefficients to a power of 2 or to the sum of a few powers of 2 . Logic is not required to multiply a tap by the coefficients $1,2,4$, and 8 -the taps are just shifted left by $0,1,2$, or 3 bits, respectively. Likewise, the numbers $3,5,6,9,10$, and 12 can be calculated with a single adder. The other numbers below 16 will require more logic, but will still be more efficient than using the conventional method.


## A

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